

# Ka-Band Power Performance of InP/InGaAs/InP Double Heterojunction Bipolar Transistors

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**Abstract**— We report for the first time the Ka-band power performance of InP/InGaAs/InP DHBT's. A  $2 \times 10 \mu\text{m}^2$  common-emitter transistor delivered a continuous wave (CW) output power of 19.1 mW (1.91 W/mm power density), an associated gain of 5.3 dB, and a power-added efficiency (PAE) of 35.5% at 30 GHz. The maximum output power density was 2.34 W/mm and the peak associated gain was 6.6 dB. Under common-base operation, the maximum associated gain increased to 15.2 dB, but the maximum output power density and peak PAE dropped to 1.91 W/mm and 24.5%, respectively, at the same frequency.

THE InP-based heterojunction bipolar transistor (HBT) technology has been progressing rapidly over the past few years. Both the current-gain cutoff frequency and maximum frequency of oscillation have broken the 200-GHz performance barrier [1]–[3]. The limitations of single HBT's (HBT's) arise from their low breakdown voltage and high output conductance, which limit their usefulness to low-voltage and low-power applications. Double HBT's (DHBT's) offer better output conductance and breakdown characteristics but require more careful base-collector heterojunction design. Too little InGaAs in the DHBT collector may lead to large saturation voltage and current blocking at low current densities. Too much InGaAs at the base-collector heterojunction results in high base-collector reverse saturation current and degradation of breakdown behavior [4]. Although no InP DHBT power amplifier has been reported to date, impressive microwave power results have been demonstrated in properly designed DHBT's with measured peak output power levels exceeding 2.0, 1.2, 0.62 W (4.3, 5.2, and 5.0 W/mm power density) at 4.3, 9, and 10 GHz, respectively, [5], [6]. We recently demonstrated that InP power DHBT's with small collector-emitter saturation voltage, large breakdown voltage and high maximum frequency of oscillation can be achieved by putting as little as 194 Å InGaAs in the >8300-Å-thick collector by carefully controlling the use of InGaAs in the collector and subcollector layers [7]. In this letter, we report on the first Ka-band power performance of these InP DHBT's.

The epitaxial structure of the InP/InGaAs/InP power DHBT consists of a 3000-Å  $n^+$  InGaAs subcollector ( $1 \times 10^{19} \text{ cm}^{-3}$ ), 8000-Å  $n^-$  InP collector ( $3 \times 10^{16} \text{ cm}^{-3}$ ), 288-Å  $n^-$  InGaAs/InP superlattice grading layer (of which 50% is InGaAs), 50-Å  $n^-$  InGaAs spacer layer, 1000-Å  $p^+$  InGaAs base ( $4 \times 10^{19} \text{ cm}^{-3}$ ), 100-Å n InGaAs setback layer ( $1 \times 10^{18}$

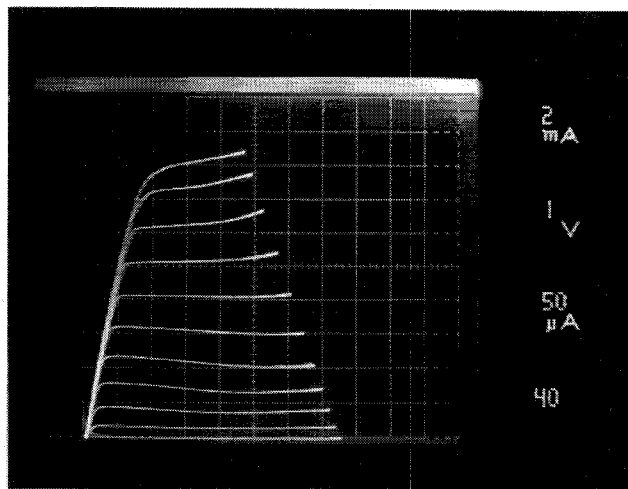


Fig. 1. Common-emitter  $I_C$ - $V_{CE}$  characteristics of a  $2 \times 10 \mu\text{m}^2$  InP/InGaAs/InP DHBT.

$\text{cm}^{-3}$ ), 900-Å n InP emitter ( $5 \times 10^{17} \text{ cm}^{-3}$ ), and 100-Å  $n^+$  InP and 1700-Å  $n^+$  InGaAs emitter contact layers (both  $1 \times 10^{19} \text{ cm}^{-3}$ ). The transistor structure was grown by metalorganic molecular beam epitaxy (MOMBE) with the use of tertiarybutylarsine (TBA) and tertiarybutylphosphine (TBP) for the arsenic and phosphorus sources, respectively, and silicon tetrabromide ( $\text{SiBr}_4$ ) and beryllium (Be) for n- and p-type doping, respectively [8].

Device fabrication began with the deposition of the emitter contact (Ti/Pt/Au), which acted as a natural mask during the base etch. The base etch was achieved by selective removal of the InGaAs cap and InP emitter layers in a phosphoric acid-based ( $\text{H}_3\text{PO}_4 : \text{H}_2\text{O}_2 : \text{H}_2\text{O}$ ) and hydrochloric acid-based ( $\text{HCl} : \text{H}_3\text{PO}_4$ ) etchants, respectively. Owing to the emitter undercut formed in the base etch, the Ti/Pt/Au base metal was self-aligned to the emitter contact without shorting the emitter finger. A reactive ion etch (RIE) process based on  $\text{CH}_4/\text{H}_2$  plasma chemistry was used to etch through the base and superlattice grading layers for the collector contact (AuGe/Ni/Au). Conventional mesa etching was used to isolate devices, and electroplated air-bridges were used to connect them to external coplanar pads and other circuit elements.

A good power transistor should exhibit high breakdown voltage, small  $V_{CE}$  offset voltage, and small  $V_{CE}$  saturation voltage. In addition, it is also important to avoid current blocking in DHBT's at low current densities. The current blocking in DHBT's is accompanied by current gain com-

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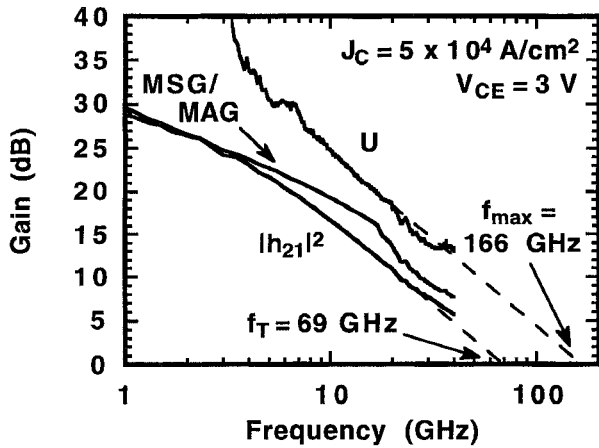


Fig. 2. Measured extrinsic RF characteristics of a  $2 \times 10 \mu\text{m}^2$  common-emitter InP/InGaAs/InP DHBT.

pression. As shown in Fig. 1, the measured common-emitter  $I_C$ - $V_{CE}$  characteristics of the completed  $2 \times 10 \mu\text{m}^2$  power DHBT's show a very small  $V_{CE}$  offset voltage of only 85 mV and a  $V_{CE}$  saturation voltage of 1 V at 10 mA collector current ( $5 \times 10^4 \text{ A/cm}^2$  current density). Current gain compression occurs at current larger than 15 mA ( $7.5 \times 10^4 \text{ A/cm}^2$  current density). In spite of the use of a total of only 194-Å InGaAs in the collector, the  $V_{CE}$  saturation voltage in the power DHBT was small. The breakdown voltages  $BV_{CBO}$  and  $BV_{CEO}$  were 29 and 18 V, respectively.

On-wafer S-parameter measurements from 0.5–40.0 GHz indicated that the transistors have a  $f_T$  and  $f_{\text{max}}$  of 69 and 166 GHz, respectively, when they were biased at  $V_{CE} = 3 \text{ V}$  and  $I_C = 12.8 \text{ mA}$  (Fig. 2). This  $f_{\text{max}}$  value is the highest in any InP DHBT that has a  $BV_{CEO}$  greater than 8 V [7]. In comparison, our Ka-band GaAs HBT's typically have a  $f_T$  and  $f_{\text{max}}$  of 41 and 107 GHz, respectively, and a  $BV_{CBO}$  and  $BV_{CEO}$  of 24 and 17 V, respectively. The maximum oscillation frequencies  $f_{\text{max}}$  were extrapolated using the  $-6 \text{ dB/octave}$  slope from the unilateral gain  $U$ . Theoretically, these extrapolated values of  $f_{\text{max}}$  should be equal to those extrapolated using  $-6 \text{ dB/octave}$  slope from the maximum available gain  $G_{\text{max}}$  [9]–[11]. Numerical simulation, however, indicated that the maximum available gain could deviate from the linear  $-6 \text{ dB/octave}$  slope as it crosses 0 dB [9]. It has also been pointed out that the best estimate of  $f_{\text{max}}$  is believed to result from the extrapolation of  $U$  because  $G_{\text{max}}$  decreases with frequency  $f$  somewhat more slowly than  $1/f^2$  [11]. The maximum available gain  $G_{\text{max}}$  of the InP DHBT's at 30 GHz was 9.3 dB, which is also considerably higher than that in our conventional Ka-band GaAs power HBT's, which have a  $G_{\text{max}}$  of about 6.9 dB at the same frequency. The results indicate that the InP HBT technology offers higher cutoff frequency, higher maximum frequency of oscillation, higher maximum available gain, and larger breakdown voltages than the corresponding GaAs HBT's.

We have measured the Ka-band power performance of the InP power transistors. Fig. 3 presents the measured common-emitter and common-base results at 30 GHz. As Fig. 3(a)

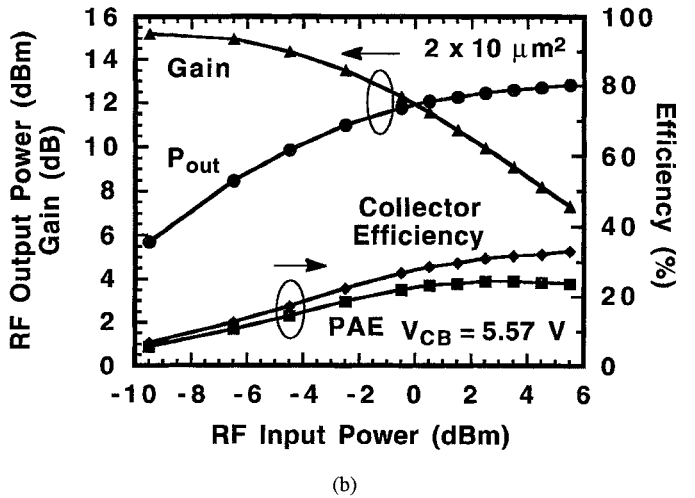
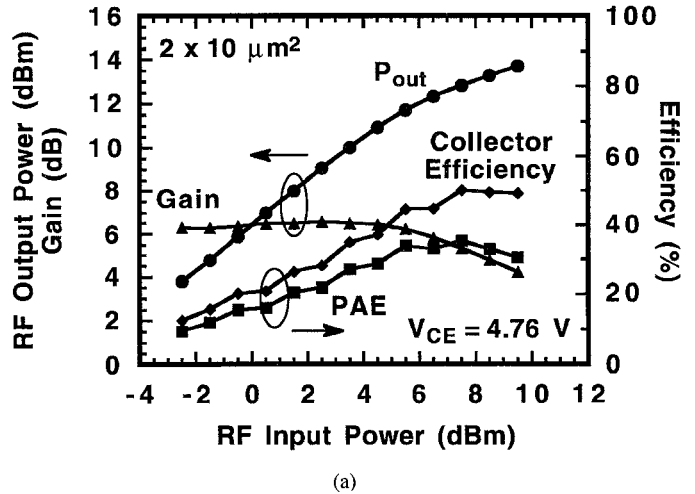


Fig. 3. Measured power performance of  $2 \times 10 \mu\text{m}^2$  InP/InGaAs/InP DHBT's at 30 GHz. (a) Common emitter. (b) Common base.

shows, the common-emitter  $2 \times 10 \mu\text{m}^2$  transistor delivered 19.1-mW CW output power (1.91-W/mm output power density), 5.3-dB gain and 35.5% power-added-efficiency (PAE) at 1-dB compression. The collector efficiency was 50.2%. The maximum output power density and the peak associated gain measured were 2.34 W/mm and 6.6 dB, respectively. The average collector current increased from 4 to 10 mA when the RF input power was increased from  $-2.5$  to 9.5 dBm. The average collector current increased with RF input power because the transistor was biased in class AB operation. Under the common-base operation, the maximum associated gain increased to 15.2 dB, but the maximum output power density and the peak PAE dropped to 1.91 W/mm and 24.5%, respectively, at the same frequency [Fig. 3(b)]. The reduced PAE in the common-base mode was due to the larger output-to-ground voltage drop ( $V_{CB}$ ) and the nonnegligible input current ( $I_E$ ). The maximum stable gain at 30 GHz was 14.2 dB in the common-base mode, but the stability factor  $k$  is less than unity throughout the entire frequency range used in the measurements. Tuning and matching were therefore found to be more difficult in the common-base transistors. In spite of the high breakdown voltage of the transistors, increasing

the  $V_{CE}$  or  $V_{CB}$  bias voltage did not result in higher output power. The maximum output power appeared to be thermally limited and the power limiting mechanisms are currently being investigated.

In summary, we reported for the first time the Ka-band power performance of InP/InGaAs/InP DHBT's. The measured extrinsic  $f_{max}$  of 166 GHz was the highest in any DHBT that has a  $BV_{CEO}$  greater than 8 V. An output power density of 1.91 W/mm with an associated gain of 5.3 dB and a PAE of 35.5% were obtained at 30 GHz for a  $2 \times 10 \mu\text{m}^2$  common-emitter DHBT. Under common-base operation, the associated gain increased to 15.2 dB, but the PAE dropped to 24.5% at the same frequency.

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